

# Discrete-Time Signal Processing

Lecture 4

z-Transforms

# The z-Transform

- The z-transform plays the same role for Discrete-Time (DT) systems that the Laplace transform plays for Continuous Time (CT) systems

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}, \quad z = \text{complex variable, } z \in \mathbb{C}$$

- The z-transform is bilateral (2-sided z-transform); unilateral z-transforms defined similarly except range on summation for  $n \geq 0$ , or  $n \leq 0$

Note:  $X(e^{j\omega}) = X(z)|_{z=e^{j\omega}}$  when it exists. Thus the discrete Fourier transform of a sequence is the z-transform evaluated on the unit circle ( $|z|=1$ ). This explains why the DTFT is periodic with period  $\omega/(2\pi)$

# Regions of Convergence (ROC)

- The Region of Convergence (ROC) of a z-transform is the region in the complex z-plane where  $|X(z)| < \infty$

$$\begin{aligned} |X(z)| &= \left| \sum_n x(n)z^{-n} \right| \leq \sum_n |x(n)| |z^{-n}| \\ &= \sum_n |x(n)| |z|^{-n} \end{aligned}$$

- $\therefore$  If  $z = z_0$  is in the ROC  $\Rightarrow$  all  $|z| = |z_0|$  will be in the ROC  
 $\Rightarrow$  ROC contains annular regions (rings centered around  $z = 0$ )
- ROC generally defined by:  $r_- < |z| < r_+$

# Regions of Convergence

Other possible ROCs:

- Note that  $z = 0$  or  $z = \infty$  may or may not be contained in these regions of convergence
- Assume  $X(z)$  is a rational function of the form:

$$X(z) = \frac{N(z)}{D(z)};$$

the roots of  $N(z)$  are the zeros of  $X(z)$  ( $X(z_0) = 0$ )

the roots of  $D(z)$  are the poles of  $X(z)$  ( $X(z_p) = \infty$ )

- Since  $X(z) \rightarrow \infty$  at the poles  $\Rightarrow$  the locations of the poles are **not** in the ROC
- The ROC of  $X(z)$  is actually bounded by the poles of  $X(z)$
- The zeros of  $X(z)$  can be either inside or outside the ROC

# Examples of ROC

- Example #1: Impulse

$$x(n) = \delta(n)$$

- Example #2: Delayed Impulse

$$x(n) = \delta(n - n_d)$$

- Example #3: Unit Step

$$x(n) = u(n)$$

# Examples of ROC

- Example #4: Causal Exponential

$$x(n) = a^n u(n) \text{ (causal)}$$

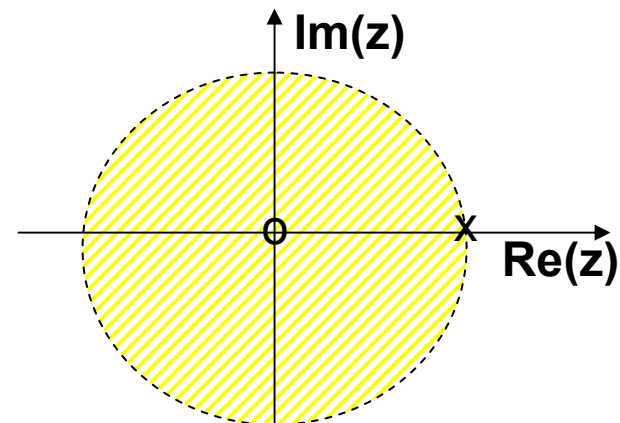
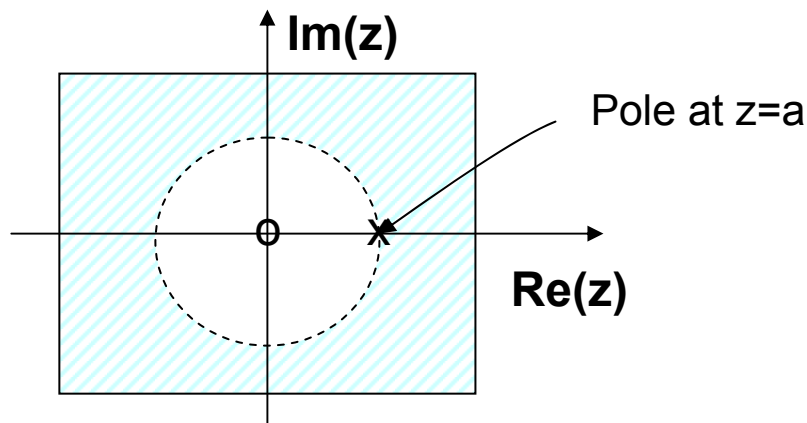
$$X(z) = \sum_{n=0}^{\infty} a^n z^{-n} = \sum_{n=0}^{\infty} (az^{-1})^n = \frac{1}{1-az^{-1}}; \text{ ROC: } |az^{-1}| < 1 \Rightarrow |z| > |a|$$

- Example #5: Non-Causal Exponential

$$x(n) = -a^n u(-n-1) \text{ (} u(-n-1) = 1; -n-1 \geq 0 \text{ or } n \leq -1; 0 \text{ otherwise)}$$

$$X(z) = \sum_{n=-\infty}^{-1} -a^n z^{-n} = -\sum_{n=-\infty}^{-1} (az^{-1})^n = \frac{1}{1-az^{-1}}; \text{ ROC: } |az^{-1}| > 1 \Rightarrow |z| < |a|$$

- Note: the functional form of the z-transforms for Examples #4 and #5 are identical, only the ROCs differ
- For causal inputs, the ROC is outside the largest pole of the z-transform
- For anti-causal inputs, the ROC is inside the smallest pole of the z-transform



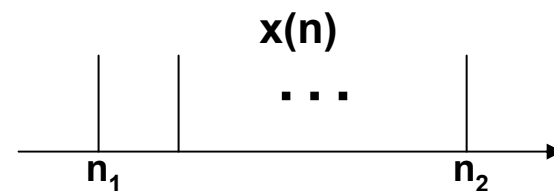
# ROC for Special Sequence Types

- Type 1: Finite Length Sequences

$$x(n) = 0, \quad n < n_1, n > n_2$$

$$X(z) = \sum_{n=n_1}^{n_2} x(n)z^{-n} = x(n_1)z^{-n_1} + \dots + x(n_2)z^{-n_2}$$

- $X(z)$  converges everywhere except possibly at  $z = 0$ , and/or  $z = \infty$ 
  - if  $n_1 \geq 0$  then ROC contains the point  $z = \infty$
  - if  $n_2 \leq 0$  then ROC contains the point  $z = 0$
- Example:  $x(n) = u(n - n_1) - u(n - n_2 - 1)$ ,  $n_2 \geq n_1$



# ROC for Special Sequence Types

- Type 2: Right-Sided Sequences

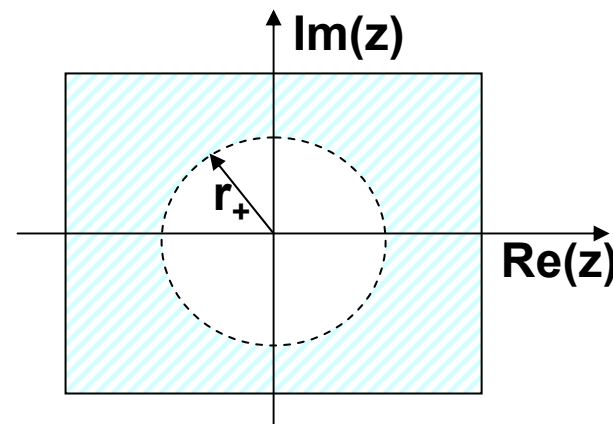
$$x(n) = 0, \quad n \leq n_0$$

$$X(z) = \sum_{n=n_0}^{\infty} x(n)z^{-n}; \quad \text{ROC: } |z| > |r_+|$$

All poles inside unit circle except possibly at  $z = \infty$

If  $n_0 \geq 0 \Rightarrow X(z)$  converges at  $z = \infty$

- Example:  $x(n) = a^n u(n+1)$ ,  $a < 1$



- Note: if  $x(n) = a^n u(n) \Rightarrow$  first term is gone and  $X(z)$  converges at  $z = \infty$

# ROC for Special Sequence Types

- Type 3: Left-Sided Sequences

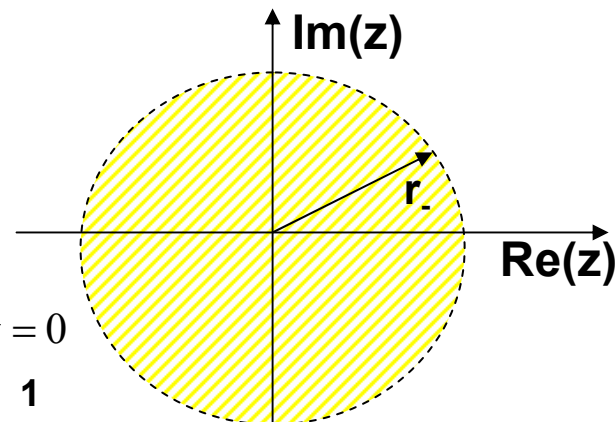
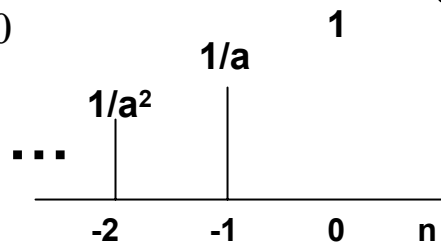
$$x(n) = 0, \quad n > n_0$$

$$X(z) = \sum_{n=-\infty}^{n_0} x(n)z^{-n}; \quad \text{ROC: } |z| < |r_-|$$

All poles outside unit circle except possibly at  $z = 0$

If  $n_0 \leq 0 \Rightarrow X(z)$  converges at  $z = 0$

- Example:  $x(n) = a^n u(-n)$ ,  $a > 1$



- Note: for this example  $n_0 = 0$  and  $X(z)$  converges at  $z = 0$

# ROC for Special Sequence Types

- Type 4: Two-Sided Sequences

$$x(n) = a^n, \quad n \geq 0 \text{ (right half), } a < 1$$

$$= b^n, \quad n < 0 \text{ (left half), } b > 1$$

$$= a^n u(n) + b^n u(-n-1)$$

$$X(z) = \sum_{n=-\infty}^{-1} b^n z^{-n} + \sum_{n=0}^{\infty} a^n z^{-n}; \quad \text{ROC: } |r_-| < |z| < |r_+|$$

- Example:  $x(n) = \left(\frac{1}{2}\right)^n u(n) + 2^n u(-n-1) = 2^{-|n|}$

Since both parts of the summation must converge  $\Rightarrow$  ROC:

# z-Transform Properties

1. ROC is a ring in the complex z-plane centered at the origin and is of the form:  $0 \leq r_- < |z| < r_+ \leq \infty$
2. The Discrete-time Fourier Transform, DTFT  $\{x(n)\}$ , of the sequence  $x(n)$  converges absolutely iff ROC contains the unit circle in the z-plane
3. The ROC does not contain any poles
4. If  $x(n)$  is a finite duration sequence, then the ROC includes the entire z-plane, except possibly the point  $z = 0$  or  $z = \infty$
5. If  $x(n)$  is a right-sided sequence, the ROC is of the form  $r_+ < |z|$ ; the ROC includes the point  $z = \infty$  if  $x(n) = 0$  for  $n < 0$
6. If  $x(n)$  is a left-sided sequence, the ROC is of the form  $r_- > |z|$ ; the ROC includes the point  $z = 0$  if  $x(n) = 0$  for  $n > 0$
7. If  $x(n)$  is a two-sided sequence, the ROC is of the form  $r_- < |z| < r_+$ ; the ROC is bounded by the poles of the z-transform of  $x(n)$
8. The ROC is always a connected region of the z-plane (no gaps or holes)

# Notes on ROC

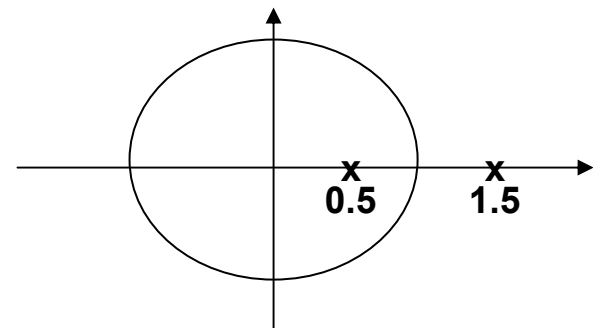
1. Must have ROC with algebraic expression to uniquely specify the z-transform--remember auxiliary conditions with LCCDES
2. Can specify system conditions such as stable system (i.e., DTFT exists  $\Rightarrow$  ROC contains the circle  $|z|=1$ ), causal system (i.e., ROC extends out from outermost finite pole), etc., rather than explicitly stating the ROC

EXAMPLE:

stable system  $\Rightarrow$  ROC:

causal system  $\Rightarrow$  ROC:

anticausal system  $\Rightarrow$  ROC:



# z-Transform Properties

$$x_1(n) \leftrightarrow X_1(z) \text{ with ROC: } R_{x_1}$$

$$x_2(n) \leftrightarrow X_2(z) \text{ with ROC: } R_{x_2}$$

## 1. Linearity:

$$ax_1(n) + bx_2(n) \leftrightarrow aX_1(z) + bX_2(z) \text{ with ROC: } \supset R_{x_1} \cap R_{x_2}$$

ROC at least the intersection of the individual ROCs;

it may be larger if there are pole-zero cancellations in the sum

## 2. Time-shifting:

$$x(n - n_d) \leftrightarrow z^{-n_d} X(z) \text{ with ROC: } R_x \text{ plus possible addition or deletion of points } z = 0 \text{ or } z = \infty$$

If  $n_d > 0$ , delaying  $x(n)$ ; if  $n_d < 0$ , advancing  $x(n)$ ; each 1 sample delay (advance) introduces a factor of  $z^{-1}$  (or  $z$ ) into the z-transform

# z-Transform Properties

## 3. Multiplication by an exponential:

$z_0^n x(n) \leftrightarrow X(z/z_0)$  with ROC:  $|z_0| R_x \Rightarrow$  all poles/zeros are multiplied by factor  $|z_0|$ , so ROC is scaled by  $|z_0|$

## 4. Differentiation:

$$nx(n) \leftrightarrow -z \frac{dX(z)}{dz} \quad \text{with ROC: } R_x$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$-z \frac{dX(z)}{dz} = -z \sum_{n=-\infty}^{\infty} -nx(n)z^{-n-1}$$

$$= \sum_{n=-\infty}^{\infty} nx(n)z^{-n} = ZT \{nx(n)\}$$

# z-Transform Properties

## 5. Complex Conjugation:

$$x^*(n) \leftrightarrow X^*(z^*) \text{ with ROC: } R_x$$

$$y(n) = x^*(n)$$

$$\begin{aligned} Y(z) &= \sum_{n=-\infty}^{\infty} x^*(n)z^{-n} = \left[ \sum_{n=-\infty}^{\infty} x(n)(z^*)^{-n} \right]^* \\ &= [X(z^*)]^* \end{aligned}$$

Note: conjugation commutes, associates, and distributes, i.e.,

$$(a+b)^* = a^* + b^*, (ab)^* = a^*b^*, (a/b)^* = a^*/b^*,$$

$$[a(b+c)]^* = a^*(b^* + c^*)$$

## 6. Time Reversal:

$$x^*(-n) \leftrightarrow X^*(1/z^*) \text{ with ROC: } 1/R_x \Rightarrow$$

if  $R_x : r_- < |z| < r_+$  then  $1/R_x : 1/r_+ < |z| < 1/r_-$

$$y(n) = x^*(-n)$$

$$\begin{aligned} Y(z) &= \sum_{n=-\infty}^{\infty} x^*(-n)z^{-n} = \left[ \sum_{m=-\infty}^{\infty} x(m)(z^{*-1})^{-m} \right]^* \\ &= X^*(z^{*-1})^* = X^*(1/z^*) \end{aligned}$$

# z-Transform Properties

## 7. Convolution of Sequences;

$$w(n) = x_1(n) * x_2(n)$$

$$W(z) = \sum_{n=-\infty}^{\infty} w(n)z^{-n} = \sum_{n=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} x_1(k)x_2(n-k)z^{-n} \text{ where ROC: } R_w$$

For  $z \in R_w$  series is absolutely convergent  $\Rightarrow$  can interchange summations, giving:

$$W(z) = \sum_{k=-\infty}^{\infty} x_1(k) \sum_{n=-\infty}^{\infty} x_2(n-k)z^{-n}$$

We recognize that the second term is just  $X_2(z)z^{-k}$  from Property #2

$$\begin{aligned} W(z) &= X_2(z) \sum_{k=-\infty}^{\infty} x_1(k)z^{-k} \\ &= X_1(z)X_2(z) \text{ with ROC: } R_w \supset R_{x_1} \cap R_{x_2} \end{aligned}$$

$R_w$  can be larger than  $R_{x_1} \cap R_{x_2}$  if there is any pole-zero cancellation

Example:

# z-Transform Properties

## 8. Initial Value Theorem:

$$x(0) = \lim_{z \rightarrow \infty} X(z) \text{ for causal } x(n) \text{ or anticausal } x(n)$$

For causal sequences we have:

since each term in the sum tends to 0 exponentially

# The Inverse z-Transform

$$x(n) = \frac{1}{2\pi j} \oint X(z)z^{n-1}dz \quad (\text{Cauchy Integral Theorem})$$

- It is hard to find  $x(n)$  by doing the integral directly!
- Instead use one of the following methods of solution:
  1. inspection
  2. properties of the z-transform
  3. partial fraction expansions
  4. long division
  5. power series expansion

# The Inverse z-Transform

## 1. Inspection:

Can use tables of z-transforms (e.g., Table 3.1 from O&S)

$$a^n u(n) \leftrightarrow \frac{1}{1 - az^{-1}}, \quad |z| > |a|$$

$$\therefore X(z) = \frac{1}{1 - 0.5z^{-1}}, \quad |z| > 0.5 \Rightarrow x(n) = (0.5)^n u(n)$$

## 2. Properties:

Use known relationships to find inverse transform

$$X(z) = \frac{z^{-3}}{1 - 0.5z^{-1}} + \frac{z^{-1}}{1 - z^{-1}}, \quad |z| > 1$$

$$x(n) = (0.5)^{n-3} u(n-3) + u(n-1)$$

# The Inverse z-Transform

## 3. Partial Fraction Expansions:

- Can use PFE to get  $X(z)$  in a form suitable for inspection or properties

$$X(z) = \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{l=0}^N a_l z^{-l}}$$

- For this function there will be  $M$  zeros and  $N$  poles at non-zero locations in the  $z$ -plane  $\Rightarrow$  can represent  $X(z)$  as

$$X(z) = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{l=1}^N (1 - d_l z^{-1})},$$

where  $c_k$  are the non-zero zeros of  $X(z)$  and  $d_l$  are the non-zero poles of  $X(z)$

- If  $M < N \Rightarrow X(z) = \sum_{k=1}^N \frac{A_k}{1 - d_k z^{-1}}, \quad A_k = (1 - d_k z^{-1}) X(z) \Big|_{z=d_k}$

# The Inverse z-Transform

Example:



# The Inverse z-Transform

Example:

# The Inverse z-Transform

## 5. Power Series:

If the z-transform is in a power series form  $\Rightarrow$  can find the coefficients of  $x(n)$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n} = \dots + x(-1)z + x(0) + x(1)z^{-1} + \dots$$

Example:  $X(z) = e^{a/z}$ ,  $|z| > 0$  (causal sequence)

$$X(z) = \sum_{n=0}^{\infty} \frac{(a/z)^n}{n!} = \sum_{n=0}^{\infty} \frac{a^n}{n!} z^{-n}$$

$$x(n) = \frac{a^n}{n!} u(n)$$

# The Inverse z-Transform

Example:  $X(z) = \frac{1}{1 - az^{-1}}, |z| > |a|, M = 0, N = 1$